A Parametric Technique for Traps Characterization in AlGaN/GaN HEMTs

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Abstract—A new parametric and cost-effective technique is developed to decouple the mechanisms behind current degradation in AlGaN/GaN HEMTs under a normal device operation: self-heating and charge trapping. A unique approach that investigates charge trapping using both source (I_S) and drain (I_D) transient currents for the first time. Two types of charge trapping mechanisms are identified: (i) bulk charge trapping occurring on a time scale of less than 1 ms, followed by (ii) surface charge trapping with a time constant larger than a millisecond. Through monitoring the difference between $I_{\rm S}$ and $I_{\rm D}$, a bulk charge trapping time constant is found to be independent of both drain $(V_{\rm DS})$ and gate $(V_{\rm GS})$ biases. Surface charge trapping is found to have a much greater impact on a slow degradation when compared to bulk trapping and self-heating. At a short timescale (< 1 ms), the RF performance is mainly restricted by both bulk charge trapping and self-heating effects. However, at a longer time (> 1 ms), the dynamic ON resistance degradation is predominantly limited by surface charge trapping.

Index Terms— AlGaN/GaN HEMTs; Transient Currents; Charge Trapping; Self-Heating Effects.

I. INTRODUCTION

AN based High Electron Mobility Transistors (HEMTs) are predicted to significantly improve efficiency and to dominate power RF applications required to deliver a high-power at a very-fast switching rate in ultra-wide-band communication, ultra-scaled radars, wireless sensors, etc. The transistor efficiency comes from wide-bandgap of III-nitrides, which have high electron saturation velocity and good thermal conductivity [1]. In recent years, these devices were steadily improving and new record performances were reported each

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year [2], [3]. However, their reliability issues persist due to the strong coupling of charge trapping and self-heating mechanisms [4]–[8].

Several studies investigating the current degradation in AlGaN/GaN HEMTs resulted in different conclusions [6], [9], [10]. Some studies of transient drain current suggest that the current degradation, at different time constants, are caused by both bulk and surface charge trapping [6], [9]. Other investigations suggest that the current degradation trends are proportional to self-heating effects that occur at different times [10]. However, a recent wide agreement is that the transient current degradation involves both device charge trapping and self-heating [11]-[14] at different times. Therefore, the time constants of the transient current need to be investigated since they are of critical consequence to device reliability and RF performance [17]. Consequently, the charge trapping is essential effect entering a design space of circuits [15] because a circuit dynamic operation affects a device operation as accounted for in an MIT virtual source GaN HEMT model [16].

This work reports the development of a parametric technique for bulk and surface charge trapping characterization in AlGaN/GaN HEMTs. The parametric technique is a new, costeffective and practical technique to acquire essential information about trap dynamics in GaN HEMTs [17], [18]. We show that a careful analysis of both the source and the drain current trends and their differences can distinguish between selfheating effects and trapping mechanisms, and determine shorttime and long-time trapping processes. Specifically, we inspect the degradation of both source and drain transient currents with the exclusion of self-heating and under normal device operation. Section II overviews the used device and experimental procedure. A unique approach that involves both source (I_S) and drain (I_D) currents to investigate the mechanisms behind transient current degradation is presented in Section III. In addition to our previously published results in [7], Section III includes (i) validation of the proposed technique for different AlGaN/GaN device architectures (gated or gateless), (ii) physical interpretations of the experimental observations through drift-diffusion simulations of the used AlGaN/GaN HEMT, and (iii) a comparison between the surface charge trapping and device self-heating effects. Conclusions are drawn in Section IV.

II. DEVICE AND EXPERIMENTAL PROCEDURE

A. Device Structure and Fabrication

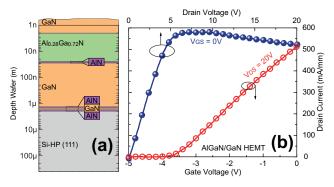


Fig. 1. (a) Schematic cross-section of the device epi-structure grown on Si-HP (111) substrate. The source-to-drain distance and the gate width are 5 μm and 100 μm , respectively. (b) $I_{\rm DS}\text{-V}_{\rm DS}$ and $I_{\rm DS}\text{-V}_{\rm GS}$ characteristics at V $_{\rm GS}$ =0 V and V $_{\rm DS}$ =20 V, respectively, obtained from the Al $_{0.28}$ Ga $_{0.72}$ N/GaN HEMT. The device threshold voltage, V $_{\rm th}$, is -3.5 V.

The epi-structure of the investigated AlGaN/GaN HEMT has been grown by molecular beam epitaxy on high-purity and high-resistivity (2000 Ω .cm) HP-Si (111) substrate. As shown in Fig. 1(a), this epi-structure consists, from the substrate to the top, of low-temperature AlN(40 nm)/GaN(250 nm)/AlN(250 nm) nucleation layers, a 1.1 μ m GaN buffer layer, a 1 nm AlN exclusion layer to reduce alloy scattering and to improve the carrier confinement of the 2-dimensional electron gas (2DEG), a 25 nm undoped Al_{0.28}Ga_{0.72}N barrier and, finally, a 1 nm undoped GaN cap layer. Room temperature Hall measurement yields a sheet resistance of 340 Ω/\Box , an electron sheet density of 1.25×10^{13} cm⁻², an electron mobility of 1480 cm $V^{-1}s^{-1}$, and a dislocation density of $\sim 5 \times 10^9$ cm⁻². The gate metallization scheme is Ni(5 nm)/Pt(25 nm)/Ti(25 nm)/Mo(30 nm)/Au(250 nm), where Ti(10 nm)/Al(200 nm)/Ni(40 nm)/Au(100 nm) multilayers have been used for the source/drain terminal. The ohmic contact resistance and specific resistivity are 0.39 Ω .mm and $3.8 \times 10^{-6} \Omega \cdot \text{cm}^2$, respectively. To reduce charge trapping effects and dispersion, the surface of the device is N2O pretreated for 2 min followed by SiO₂(100 nm)/Si₃N₄(50 nm) bi-layer passivation. The fabrication process is similar to that described in Ref. [19]. The current difference of I_{DS}- $V_{\rm GS}$ at $V_{\rm DS}$ =20 V and the current difference of $I_{\rm DS}\text{-}V_{\rm DS}$ at V_{GS} =0 V characteristics of the AlGaN/GaN HEMT are given in Fig. 1(b). These electrical characteristics are performed at DC and dark conditions using Agilent B1500A framework [7].

B. Experimental Methodology

For the first time, both source and drain transient currents are measured to investigate charge trapping, illustrated in Fig. 2(a), in AlGaN/GaN HEMTs under normal device operation. The pulse waveforms of the first experiment, given in Fig. 2(b), are used to characterize the transient current degradation mechanism. Here, the drain voltage $V_{\rm DS}=0~V$ and $V_{\rm GS}=0~V$ are pulsed to $V_{\rm DS1}$ and $V_{\rm GS1}$, respectively,

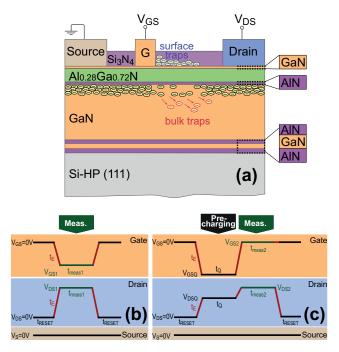


Fig. 2. (a) Illustration of bulk and surface charge trapping mechanisms in an AlGaN/GaN HEMT at a semi—ON state. Pulse waveforms used for measurements of transient source and drain currents, $I_{\rm S}$ and $I_{\rm D}$, are for: (b) without traps pre-charging and (c) with traps pre-charging phase. The used refresh/reset time, $t_{\rm RESET}$, is 10 s and all rising/falling time edges are set to time $t_{\rm E}$ =200 ns.

for a measurement time of $t_{\rm meas1}=1$ s. The pulse waveforms of the second experiment, shown in Fig. 2(c), are similar to the first experiment but involve a pre-charging phase with quiescent biasing conditions, $V_{\rm DSQ}$ and $V_{\rm GSQ}$. Measurements are then taken at $V_{\rm DS2}=10$ V and $V_{\rm GS2}=0$ V for a time $t_{\rm meas2}=1$ s. The corresponding results of these two experiments (schematically illustrated in Fig. 2(b) and (c)) are presented and analysed in Section III.

In all experiment, the time to refresh/reset the device is set to $t_{\rm RESET}=10~\rm s$ to recover the device from trapped charges and self-heating that would affect further measurements. All rising/falling time edges are fixed to $t_{\rm E}=200~\rm ns$. The trace of the transient current is reproducible revealing that no permanent degradation of the current occurs from the experiment, only recoverable degradation, i.e., charge trapping and self-heating (not shown).

III. TRANSIENT CURRENT DEGRADATION

Using the experimental methodology of Fig. 2(b), $I_{\rm S}$ and $I_{\rm D}$, transient currents are measured on both gated and gateless AlGaN/GaN devices as shown in Fig. 3(a) and Fig. 3(b), respectively. Similar trends of degradation phenomena persist regardless of the device architecture, gated or gateless. The transient behaviour of $I_{\rm S}$ and $I_{\rm D}$ can be broadly split into two phases, a fast degradation (DG1) for time windows shorter than a millisecond, and a slow degradation (DG2) for a longer time (> 1 ms). Comparable to the gated device, $I_{\rm S}$ - $I_{\rm D}$ > 0 mA mm⁻¹ occurred in the gateless device. This indicates that a difference between $I_{\rm S}$ and $I_{\rm D}$ is not caused by the gate terminal. Therefore, the later described analysis

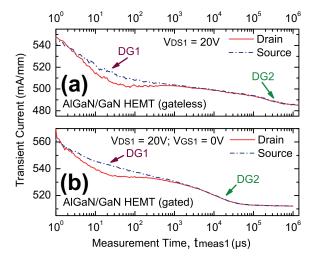


Fig. 3. Transient source and drain currents, I_S and I_D , of (a) a gateless AlGaN/GaN HEMT and (b) a gated AlGaN/GaN HEMT. Very similar degradation mechanisms, one fast (DG1) and one slow (DG2), are observed in both devices. The experimental conditions and the used pulse waveforms are summarized in Fig. 2(b).

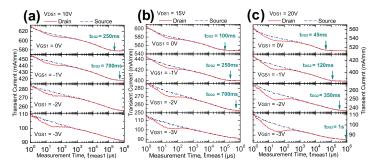


Fig. 4. Transient behaviour of the drain, $\rm I_D$, and the source, $\rm I_S$, currents versus the measurement time, $t_{\rm meas1}$, on log scale using the pulse waveforms given in Fig. 2(b) at $\rm V_{GS1}=0$ V to -3 V with (a) $\rm V_{DS1}=10$ V and (b) $\rm V_{DS1}=20$ V.

of these current transients is not only limited to AlGaN/GaN HEMTs.

In the following subsections, we investigate the mechanisms of DG1 and DG2 under various biasing conditions. Using the same pulse waveforms given in Fig. 2(b), we apply $V_{\rm DS1} = 10~V$, 15 V and 20 V, and $V_{\rm GS1} = 0~V$ to -3 V (step -1 V) to the AlGaN/GaN HEMT. These conditions place the device in a semi–ON state. The transient currents for each gate biasing condition are shown in Fig. 4(a) for $V_{\rm DS1} = 10~V$, and in Fig. 4(b) for $V_{\rm DS1} = 20~V$.

A. Fast Current Degradation - DG1

During the fast current degradation phase, DG1, a significant magnitude of $I_{\rm S}\text{-}I_{\rm D}$ is observed. Electrons that are injected from the source terminal and get trapped within the bulk, e.g. the GaN buffer, are not collected by the drain terminal, resulting in a difference between $I_{\rm S}$ and $I_{\rm D}$ ($I_{\rm S}\text{-}I_{\rm D}>0$ mA mm $^{-1}$). This stems from the applied electric field that provides enough energy for charge carriers to be trapped within the bulk [9]. This fast degradation saturates within the first 100 μs . The self-heating and trapped surface charges do not influence this difference, $I_{\rm S}\text{-}I_{\rm D}$, as they degrade both $I_{\rm S}$

and I_D proportionally [5], [20]. The mechanism behind this fast current degradation has to contribute to the bulk charge trapping and self-heating.

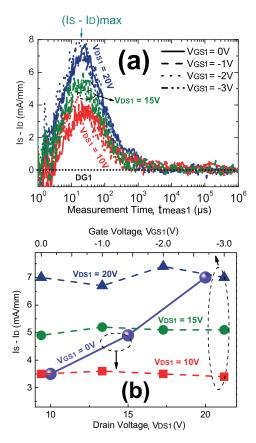


Fig. 5. (a) I_S and I_D difference (I_S-I_D) versus the measurement time at V_{DS1} = 10 V, 15 V and 20 V for different gate voltages (V_{GS1} = 0 V to -3 V); indicating the bulk charge trapping process (DG1). I_S-I_D timing characteristic is independent of both drain and gate biases. (b) The impact of drain voltage, V_{DS1}, and gate voltage, V_{GS1}, on I_S-I_D showing a dependence of the bulk trapped charge density and V_{DS1}. Unlike V_{DS1}, V_{GS1} shows a negligible impact on bulk trapping characteristics.

Due to applied electric field between the gate and the drain, charge carriers can be trapped at the surface of the device where dangling bonds create surface states, as illustrated in Fig. 2(a). As the surface charge trapping occurs within a short period, typically within the first microsecond [21], its contribution towards the DG1 is difficult to distinguish from self-heating.

To analyse the dependence of the bulk charge trapping on drain and gate voltages, Fig. 5(a) shows $I_{\rm S}\text{-}I_{\rm D}$ values with respect to measurement time. From these measurements, one can deduce that the time constant, $t_{\rm DG1}$, of the DG1 is independent of biasing conditions. This suggests that the rate/speed of bulk charge trapping does not change as this is dependent of the density of threading dislocations and defects [9].

The magnitude of the current drop due to trapped bulk charges, I_S - I_D value, taken from Fig. 5(a) at measurement time ($t_{\rm meas1} \approx 35~\mu s$), is plotted against each drain and gate biasing condition in Fig. 5(b). We observe an increase of bulk charge trapping with respect to the drain voltage,

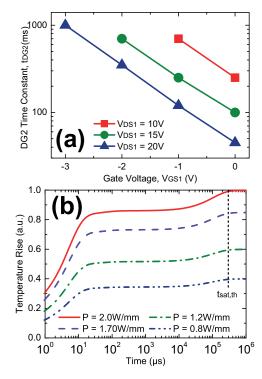


Fig. 6. (a) Time constant of DG2, $t_{\mathbf{DG2}}$, at $V_{\mathbf{DS1}}=10$ V, 15 V, and 20 V and $V_{\mathbf{GS1}}=0$ V to -3 V; revealing the dependence of $t_{\mathbf{DG2}}$ on both drain and gate voltages. (b) Transient heating characteristics at various power densities (P) using the RC thermal model [10], [25]. Note that $t_{\mathbf{sat,th}}$ is temperature independent.

 $V_{\rm DS1}$. Increasing the drain bias, $V_{\rm DS1}$, provides more kinetic energy to carriers to reach deeper traps in the access region that results in an increase of the bulk trapping. The gate bias $V_{\rm GS1}$ has, however, insignificant impact on the bulk charge trapping. Further experiments are given in Section III-C to investigate the relationship between the bias conditions and the bulk charge trapping.

B. Slow Current Degradation - DG2

Throughout the slow current degradation phase, DG2, current begins to diminish a second time, after the saturation of DG1 and the first phase of self-heating. A negligible $I_{\rm S}\text{-}I_{\rm D}$ magnitude is observed indicating an insignificant bulk charge trapping in this phase of the DG2. Unlike bulk charge trapping, where $I_{\rm S}\text{-}I_{\rm D}>0$ mA mm $^{-1}$, the surface charge trapping and self-heating alter the electrostatic integrity and channel resistance in the device. As a result, both source and drain currents degrade proportionally $(I_{\rm S}\text{-}I_{\rm D}=0$ mA mm $^{-1})$.

We extracted the time constant of DG2 ($t_{\rm DG2}$) from Fig. 4 and presented it in Fig. 6(a) with respect to both drain and gate biasing conditions. The increase of $V_{\rm GS1}$ induces a greater surface trapping density, leading to a larger time for the surface charges to redistribute between gate and drain terminals. However, the required time to redistribute the trapped charges at the surface and to extend the 'virtual gate' towards the drain side reduces with increased $V_{\rm DS1}$. A larger drain voltage induces a higher electric field and, therefore, distributes surface charges more quickly [22]. The extracted $t_{\rm DG2}$ corresponds with that of other works [23], [24].

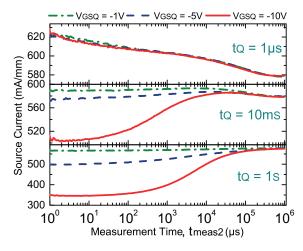


Fig. 7. Transient behaviour of the source current, I_S , versus the measurement time, t_{meas2} , using the pulse waveforms given in Fig. 2(c) with $V_{DSQ} = V_{DS2} = 10$ V, $V_{GSQ} = -1$ V, -5 V, and -10 V, and $V_{GS2} = 0$ V. The effect of current recovery is shown to increase with greater $|V_{GSQ}|$ and t_Q due to a greater surface trapping density and re-distribution.

In contrast to the time constant of DG1, $t_{\rm DG2}$ is both drain and gate bias dependent. The slow degradation of DG2 can also be coupled with a possible second phase of self-heating. Evidence of this second phase of self-heating comes from the RC thermal model shown in Fig. 6(b) [10], [25]. We observe that the time constant of this second transient heating ($t_{\rm sat,th}$) is independent of the applied power. Therefore, the surface trapping is the cause of $t_{\rm DG2}$ variation.

The pulse waveforms given in Fig. 2(c) are used to investigate the dominant degradation mechanism of the DG2. Precharging quiescent conditions, $V_{\rm DSQ}=10~\rm V$ and $V_{\rm GSQ}=-10~\rm V$, $-8~\rm V$, $-5~\rm V$, $-3~\rm V$, and $-1~\rm V$, are set at precharging times $t_{\rm Q}=1~\mu \rm s$, $10~\rm ms$, and $1~\rm s$. After the device is pre-charged, the transient current measurements are taken at $V_{\rm DS2}=10~\rm V$ and $V_{\rm GS2}=0~\rm V$ for a time of $t_{\rm meas}$, $2~\rm from$ $1~\mu \rm s$ to $1~\rm s$. The corresponding results of this experiment are shown in Fig. 7.

During the pre-charging phase, surface charge trapping and distribution occur and are controlled by $V_{\rm GSQ}$ and $V_{\rm DSQ}.$ No device self-heating effects can take place when the device is OFF ($V_{\rm GSQ} < V_{\rm th}$). However, when the device is in a semi–ON state ($V_{\rm GSQ} > \! V_{\rm th}$), a self-heating will occur in addition to the bulk charge trapping due to a high $V_{\rm DSQ}$ of 10 V.

Through the measurement phase, the trapped surface charges begin to re-distribute in the opposite manner to the precharging conditions and surface trapping density reduces. This is the consequence of the decreased magnitude of gate voltage from the quiescent to the measurement condition. Self-heating increases as the device is switched from the OFF/semi–ON state to the ON–state. For all quiescent bias conditions where $t_{\rm Q}>1$ ms, bulk charge trapping is saturated during the precharging phase at the same amount as found in Section III.A and no further bulk trapping takes place in the measurement phase. The behaviour of transient current is analyzed for each pre-charging condition:

• $t_Q = 1 \mu s$ - For all V_{GSQ} , negligible surface/bulk charge

trapping and self-heating are induced during the precharging condition because of short pre-charging time. As a result, no recovery of surface trapping occurs during the measurement. Instead, bulk trapping, self-heating, and surface trapping degrade the current. Therefore, the transient current behaviour is similar to that of Fig. 4(a) at $V_{\rm GS1}=0$ V.

- $t_Q = 10$ ms During the pre-charging phase, surface charge trapping/distribution and complete bulk charge trapping occur. No self-heating is induced during the precharging for $V_{\rm GSQ}$ =-5 V and -10 V. As a result, the primary cause of the degraded current at $t_{meas,2} = 1 \mu s$ is the surface and bulk charge trapping. This initial degradation increases with larger |V_{GSQ}| as a larger surface charge density is trapped/distributed. A recovery of the current is then observed during measurement from $\sim 10 \ \mu s$ to ~ 40 ms at $V_{\rm GSQ}$ =-1 V to -10 V, respectively. This shows that the surface trapping recovery has a greater impact on current behaviour than the degradation of the current due to self-heating. Since the time $t_0 = 10 \text{ ms}$ is not sufficiently long to complete the surface trapping process during the pre-charging phase, as observed in Fig. 4(a), the surface trapping recovery ends before the current degradation is complete during the measurement phase. Hence, a degradation of the current is observed beyond ~ 40 ms as the result of self-heating effects.
- $t_{\rm Q}=1~{\rm s}$ Pre-charging of surface traps is complete and at a larger magnitude. The current recovers throughout the entire measurement window. A large recovery of the current is observed due to the dominance of a recovery from the surface traps over a recovery from self-heating.

Fig. 7 illustrates that, at $t_Q \geq 10$ ms, two mechanisms of transient current develop when reducing the gate voltage from $|V_{\rm GSQ}|$ to $|V_{\rm GS2}|$: (i) a recovery of surface trapping redistribution whereby a 'virtual gate' length is reduced, lowering the channel resistance and recovering the current, and (ii) a self-heating that increases the channel resistance and degrades the current.

C. Bulk charge trapping versus bias conditions

In this subsection, we investigate the relationship between the bulk charge trapping and the bias conditions. Contrary to the previous experiment of Fig. 7, the bulk charge trapping is excluded from the pre-charging phase whilst maintaining the surface trapping throughout the pre-charging. To achieve this, $V_{DSQ} = 0$ V is used (Fig. 2(c)). The other pre-charging conditions are $V_{\rm GSQ}$ =-1 V, -3 V, -8 V and -10 V for $t_{\rm Q}$ =1 s. The measurement settings are $V_{\rm GS2}$ =0 V and $t_{\rm meas,2}$ =1 s with different drain voltages of $V_{\rm DS2}$ =10 V and 20 V, as shown in Fig. 8(a) and Fig. 8(b), respectively. The difference between I_S and I_D is extracted from Fig. 8 and presented in Fig. 9 showing near identical characteristics to those in Fig. 5(a). These nearly identical characteristics confirm that (i) no bulk trapping had taken place during the pre-charging at $V_{DSQ} = 0$ V and (ii) bulk trapping rate is not affected by the gate bias during the ON and OFF states of the devices, only by the drain voltage.

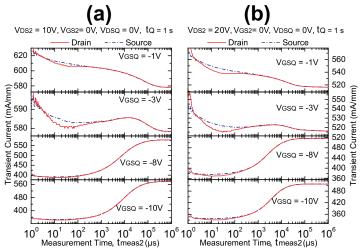


Fig. 8. Transient behaviour of the drain, I_D , and the source, I_S , currents, versus the measurement time, $t_{\rm meas2}$, on log scale using the pulse waveforms given in Fig. 2(c) at different pre-charging conditions ($V_{\rm DSQ}=0$ V; $V_{\rm GSQ}=-1$ V, -3 V, -8 V, and -10 V; $t_{\rm Q}=1$ s) with (a) $V_{\rm DS2}=10$ V; $V_{\rm GS2}=0$ V and (b) $V_{\rm DS2}=20$ V; $V_{\rm GS2}=0$ V. The bulk trapping has a minimal impact on the recovery of current degradation when pre-charging conditions set the device to the OFF-state; $|V_{\rm GSQ}|>V_{\rm th}$.

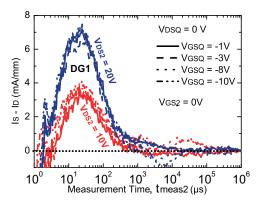


Fig. 9. I_{S} and I_{D} difference (I_{S} - I_{D}) versus the measurement time at a gate voltage V_{GS2} =-1 V and two drain voltages (V_{DS2} =10 V and 20 V) showing that charge carrier trapping in the bulk is negligible during pre-charging phase when compared with Fig. 5(a).

The electric field distribution along the channel is extracted at the AlN/GaN interface, at different $V_{\rm GS}$ and $V_{\rm DS}$ bias conditions, from drift-diffusion simulations of the device described in Section II.A. At a constant V_{DS}, the increase of the electric field with V_{GS} at the drain edge of the gate is related to a drop of the field in the access region as shown in Fig. 10(a) while, at a constant V_{GS} , the electric field increases with $V_{\rm DS}$ in the whole channel (Fig. 10(b)). Therefore, to assess the trapping process over the whole channel and not just in a particular area of the device, we have calculated the overall electric field for all bias conditions by averaging the electric field along the transistor channel as shown in Fig. 10(c). The figure illustrates that the increase in $V_{\rm DS}$ at a fixed V_{GS} increases the overall electric field linearly (the bottom scale of Fig. 10(c)), while the decrease in V_{GS} at a fixed V_{DS} keeps the overall electric field practically constant (the top scale of Fig. 10(c)) [26]. The average electric field as a

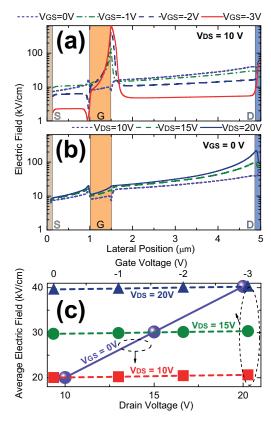


Fig. 10. Drift-diffusion simulation of electric field distributions in the channel of the AlGaN/GaN HEMT: (a) at $V_{\rm DS}=10$ V and different $V_{\rm GS}$ and (b) at $V_{\rm GS}=0$ V and different $V_{\rm DS}$. (c) Comparison of the average electric field at different bias conditions showing a negligible impact of the gate voltage on the overall electric field.

function of the gate and the drain applied biases, presented in Fig. 10(c), are thus in a good correlation with experimental data shown in Fig. 5(b).

The details of the used device simulation technique can be found in [4], [5], [27]–[30]. In the past, this simulation technique has been successfully used to predict performance of AlGaN/GaN HEMTs grown on various substrates (i.e., Si, 4H-SiC, diamond, sapphire). The transport parameters such as electron drift velocity, energy relaxation time, and electron effective mass are obtained from Monte Carlo simulations at different lattice temperatures [30]. Both simulated I-V characteristics and calculated temperatures have been compared to experimental I-V characteristics and temperature measurements, demonstrating a good agreement [5], [27]–[29]. This simulation technique has been adapted for both gated and gateless AlGaN/GaN HEMTs.

IV. CONCLUSIONS

A new source and drain transient currents (I_S and I_D) technique for a charge trapping characterization in AlGaN/GaN HEMTs, under normal device operation, has been developed. This proposed technique takes into account the device operating temperature variation during the transient measurements, a cost-effective complement to the temperature-dependent current-transient based techniques [6], [9], [23], [31]–[34]. It can be easily implemented to decouple the slow bulk charge

traps and the surface traps from self-heating effects without turning to device simulations or to more complicated measurements. Using this technique, charge trapping behaviour with the exclusion of self-heating has been analyzed. Two types of charge trapping mechanisms have been identified: (i) a bulk charge trapping occurring on a time scale of less than 1 ms, followed by (ii) a surface charge trapping and distribution beyond 1 ms. The bulk trapping and the surface trapping corresponds to the primary fast and slow current degradation mechanisms, respectively.

Through monitoring the difference between $I_{\rm S}$ and $I_{\rm D}$, the bulk charge trapping time constant is shown to be independent of both $V_{\rm DS}$ and $V_{\rm GS}$, although, $V_{\rm DS}$ is found to affect the bulk trap density. Large $V_{\rm DS}$ is found to be a cause of the bulk charge trapping during both the ON and the semi-ON states of the device. The surface trapping is found to have a much larger impact on the slow degradation of the current when compared to the self-heating and the bulk charge trapping in the investigated device structures. This is an important step to understand the priority of the device engineering, whereby the focus should be aimed towards reducing the surface trapping in order to improve an RF performance and a dynamic ON resistance [8].

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